

SNS COLLEGE OF TECHNOLOGY



An Autonomous Institution Coimbatore-35

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DEPARTMENT OF ELECTRONICS & COMMUNICATION ENGINEERING

19ECB212 - DIGITAL SIGNAL PROCESSING

II YEAR/ IV SEMESTER

UNIT 2 – IIR FILTER DESIGN

TOPIC - BUTTERWORTH FILTER



DESIGN OF LOWPASS DIGITAL BUTTERWORTH FILTER



- The popular methods of designing IIR digital filter involves the design of equivalent analog filter and then converting the analog filter into digital filter
- Hence to design a Butterworth IIR digital filter, first an analog butterworth filter transfer function is determined using the given specifications
- Then the analog filter transfer function is converted to a digital filter transfer function by using either Impulse Invariant Transformation (or) Bilinear Transformation



ANALOG BUTTERWORTH FILTER



- The analog Butterworth filter is designed by approximating the ideal analog filter frequency response, $H(j\Omega)$ using an error function
- The error function is selected such that the magnitude is maximally flat in the passband and monotonically decreasing in the stopband (The magnitude is maximally flat at the origin i.e., Ω = 0 and monotonically decreasing with increasing Ω
- The magnitude response of lowpass filter obtained by this approximation is given by $|\mathbf{H}(\Omega)|^2 = \frac{1}{1+\left[\frac{\Omega}{1}\right]^{2N}}$



PROPERTIES OF BUTTERWORTH FILTERS



- The Butterworth filters are all pole designs (i.e., the zeros of the filters exist at infinity)
- At the cutoff frequency Ω_c the magnitude of normalized Butterworth filter is $1/\sqrt{2}$ (i.e., $|H(j\Omega)| = 1/\sqrt{2} = 0.707$) Hence the dB magnitude at the cutoff frequency will be 3 dB less than the maximum value
- The filter order N completely specifies the filter
- The magnitude is maximally flat at the origin
- The magnitude is a monotonically decreasing function of Ω
- The magnitude response approaches the ideal response as the value of N increases



TRANSFER FUNCTION OF ANALOG BUTTERWORTH LOWPASS FILTER



- For a stable and causal filter the poles should lie on the left half of s-plane.
 Hence the digital filter transfer function is formed by choosing the N number of left half poles
- When N is even, all the poles are complex and exist as conjugate pair. When
 N is odd, one of the poles is real and all other poles are complex and exist
 as conjugate pair
- Therefore the transfer function of Butterworth filters will be a product of second order factors



NORMALIZED BUTTERWORTH LPF TRANSFER FUNCTION



- N be the order of the filter
- H(s_n) be the normalized Butterworth lowpass filter function
- When N is even,

$$H(s_n) = \prod_{k=1}^{\frac{N}{2}} \frac{1}{s_n^2 + b_k s_n + 1}$$

When N is odd,

$$H(s_n) = \frac{1}{s_n + 1} \frac{\prod_{k=1}^{N-1} \frac{1}{s_n^2 + b_k s_n + 1}}{where, b_k = 2 \sin\left[\frac{(2k-1)\pi}{2N}\right]}$$



UNNORMALIZED BUTTERWORTH LPF TRANSFER FUNCTION



- The unnormalized transfer function is obtained by replacing s_n by s/Ω_c in the normalized transfer function, where Ω_c is the 3 dB cutoff frequency of the lowpass filter
- H(s) be the normalized Butterworth lowpass filter function
- When N is even,

$$\therefore H(s) = \prod_{k=1}^{\frac{N}{2}} \frac{1}{s_n^2 + b_k s_n + 1} \Big|_{s_n = \frac{s}{\Omega_c}}$$

$$= \prod_{k=1}^{\frac{N}{2}} \frac{\Omega_c^2}{s^2 + b_k \Omega_c s + \Omega_c^2}$$



UNNORMALIZED BUTTERWORTH LPF TRANSFER FUNCTION



- H(s) be the normalized Butterworth lowpass filter function
- When N is odd, H(s) is obtained by letting $s_n\to s/\Omega_c$ in the normalized Butterworth lowpass filter function

$$\therefore H(s) = \frac{1}{s_n + 1} \frac{\prod_{k=1}^{N-1}}{\prod_{k=1}^{2} \frac{1}{s_n^2 + b_k s_n + 1}} \Big|_{s_n = \frac{s}{\Omega_c}}$$

$$= \frac{\Omega_c}{s + \Omega_c} \frac{\prod_{k=1}^{N-1}}{\prod_{k=1}^{2} \frac{\Omega_c^2}{s^2 + b_k \Omega_c + \Omega_c^2}}$$



BUTTERWORTH LPF NORMALIZED TRANSFER FUNCTION

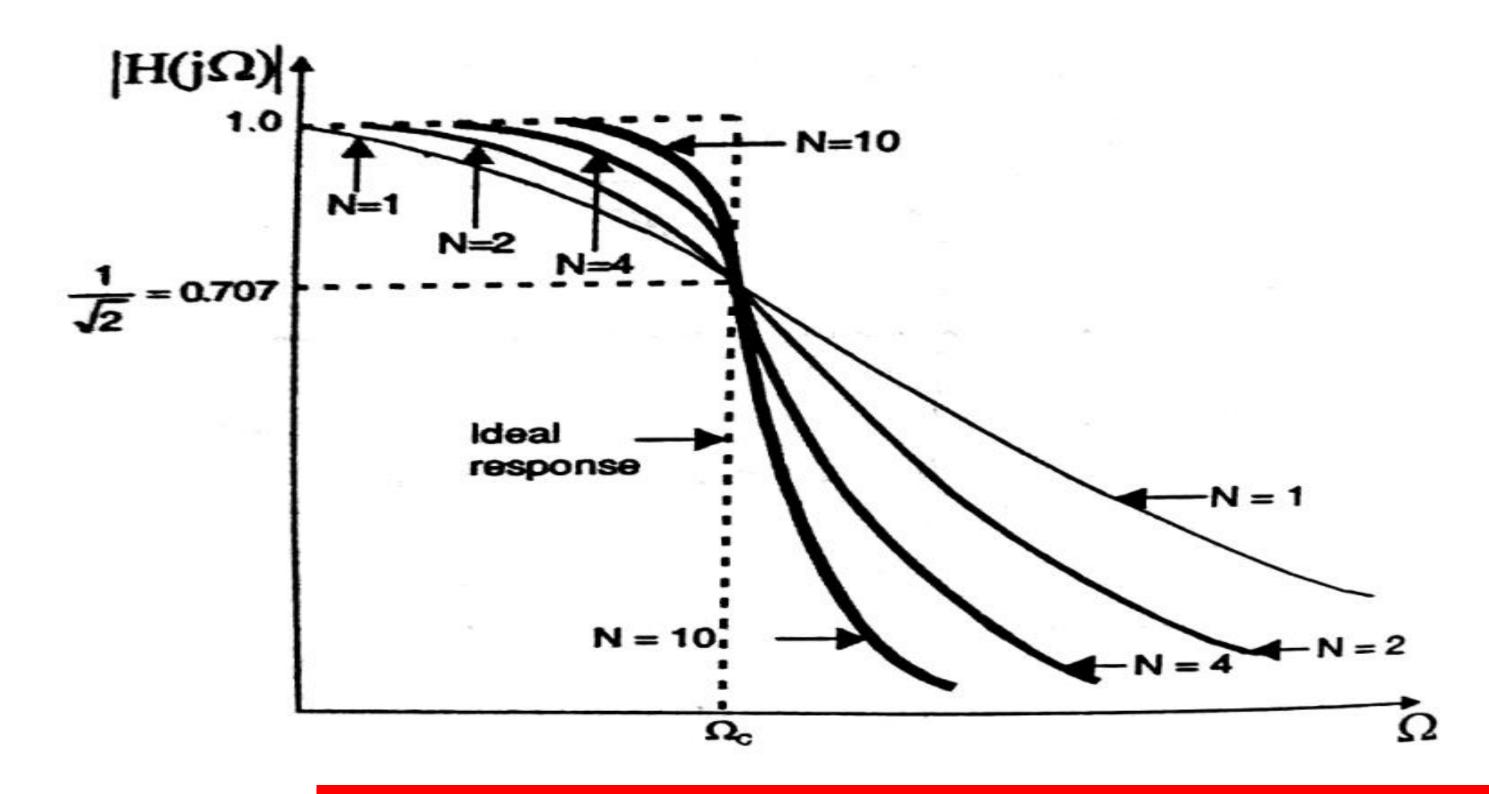


Order, N	Normalized tansfer function, H(s _n)
1	$\frac{1}{s_n+1}$
2	$\frac{1}{s_n^2 + 1.414 s_n + 1}$
3	$\frac{1}{(s_n + 1) (s_n^2 + s_n + 1)}$
4	$\frac{1}{(s_n^2 + 0.765s_n + 1)(s_n^2 + 1.848s_n + 1)}$
5	$\frac{1}{(s_n + 1) (s_n^2 + 0.618s_n + 1) (s_n^2 + 1.618s_n + 1)}$
6	$\frac{1}{(s_n^2 + 1.932 s_n + 1) (s_n^2 + 1.414 s_n + 1) (s_n^2 + 0.518 s_n + 1)}$



FREQUENCY RESPONSE OF ANALOG LOWPASS BUTTERWORTH FILTER







ORDER OF THE LOWPASS BUTTERWORTH FILTER



- In Butterworth filters the frequency response of the filter depends on the order N.
 The specifications of the filter are given in terms of gain at a passband and stopband frequency
- ${\bf A_p}$ Gain or Magnitude at pass band edge frequency ${\bf \Omega_p}$
- \mathbf{A}_s Gain or Magnitude at Stop band edge frequency $\mathbf{\Omega}_s$

$$N_{1} = \frac{1}{2} \frac{\log \left[\frac{\left(1/A_{s}^{2} \right) - 1}{\left(1/A_{p}^{2} \right) - 1} \right]}{\log \left(\frac{\Omega_{s}}{\Omega_{p}} \right)}$$



ORDER OF THE LOWPASS BUTTERWORTH FILTER



- The specifications of the filter are given in terms of dB attenuation at a passband and stopband frequency
- $\alpha_{p,dB}$ dB attenuation at pass band edge frequency Ω_p
- $\alpha_{s,dB}$ dB attenuation at Stop band edge frequency Ω_s

$$N_{1} = \frac{\log \left[\left(\frac{10^{0.1\alpha_{s,dB}} - 1}{10^{0.1\alpha_{p,dB}} - 1} \right)^{\frac{1}{2}} \right]}{\log \frac{\Omega_{s}}{\Omega_{p}}}$$



LOWPASS BUTTERWORTH FILTER



Bilinear Transformation:

$$\Omega_{\rm p} = \frac{2}{T} \tan \frac{\omega_{\rm p}}{2}$$

$$\Omega_{\rm s} = \frac{2}{T} \tan \frac{\omega_{\rm s}}{2}$$

• Impulse Invariant Transformation:

$$\Omega_{\mathbf{p}} = \frac{\omega_{\mathbf{p}}}{T}$$

$$\Omega_{\rm s} = \frac{\omega_{\rm s}}{T}$$



CUTOFF FREQUENCY OF LOWPASS BUTTERWORTH FILTER



• When the specifications are A_p , A_s , ω_p , ω_s

Cutoff frequency,
$$\Omega_c = \frac{\Omega_s}{\left[\left(1/A_s^2\right) - 1\right]^{\frac{1}{2N}}}$$

Cutoff frequency,
$$\Omega_c = \frac{32_p}{\left[\left(1/A_p^2\right)-1\right]^{\frac{1}{2N}}}$$



CUTOFF FREQUENCY OF LOWPASS BUTTERWORTH FILTER



• When the specifications are $\alpha_{p,\,dB}$, $\alpha_{s,\,dB}$, ω_p , ω_s

Cutoff frequency,
$$\Omega_{c} = \frac{\Omega_{s}}{\left(10^{0.1\alpha_{s,dB}} - 1\right)^{\frac{1}{2N}}}$$
Cutoff frequency, $\Omega_{c} = \frac{\Omega_{p}}{\left(10^{0.1\alpha_{p,dB}} - 1\right)^{\frac{1}{2N}}}$



DESIGN PROCEDURE FOR LOWPASS DIGITAL BUTTERWORTH IIR FILTER



- ω_p Pass band edge digital frequency in rad /sample
- ω_s Stop band edge digital frequency in rad/sample
- A_p Gain at pass band edge frequency ω_p
- A_s Gain at Stop band edge frequency ω_s
- $T = 1/F_s$ Sampling time in sec.
- Where F_s = Sampling frequency in Hz
- Ω_p Pass band edge analog frequency corresponding to ω_p
- Ω_s Stop band edge analog frequency corresponding to ω_s



DESIGN PROCEDURE FOR LOWPASS DIGITAL BUTTERWORTH IIR FILTER



- 1. Choose either Bilinear or Impulse Invariant transformation and determine the specifications of equivalent analog filter
- The gain or attenuation of analog filter is same as digital filter
- Bilinear Transformation:

$$\Omega_{\rm p} = \frac{2}{\rm T} \tan \frac{\omega_{\rm p}}{2}$$

$$\Omega_{\rm s} = \frac{2}{T} \tan \frac{\omega_{\rm s}}{2}$$

• Impulse Invariant Transformation:

$$\Omega_{\mathbf{p}} = \frac{\omega_{\mathbf{p}}}{T}$$

$$\Omega_{\rm s} = \frac{\omega_{\rm s}}{T}$$



ORDER OF THE LOWPASS DIGITAL BUTTERWORTH FILTER



2. Decide the order N of the filter. In order to estimate the order N, Calculate the Parameter N_1 using the following equation:

$$N_{1} = \frac{1}{2} \frac{\log \left[\frac{\left(1/A_{s}^{2} \right) - 1}{\left(1/A_{p}^{2} \right) - 1} \right]}{\log \left(\frac{\Omega_{s}}{\Omega_{p}} \right)}$$

• Choose N such that, N \geq N₁, Usually N is chosen as nearest integer just greater than N₁



NORMALIZED BUTTERWORTH LPF TRANSFER FUNCTION



- 3. Determine the normalized transfer function $H(s_n)$ of the analog lowpass filter function
- When N is even,

$$H(s_n) = \prod_{k=1}^{\frac{N}{2}} \frac{1}{s_n^2 + b_k s_n + 1}$$

When N is odd,

$$H(s_n) = \frac{1}{s_n + 1} \frac{\prod_{k=1}^{N-1} \frac{1}{s_n^2 + b_k s_n + 1}}{where, b_k = 2 \sin\left[\frac{(2k-1)\pi}{2N}\right]}$$



CUTOFF FREQUENCY



4. Calculate the analog Cutoff frequency $\Omega_{\rm c}$

Cutoff frequency,
$$\Omega_c = \frac{32_s}{\left[\left(1/A_s^2\right)-1\right]^{\frac{1}{2N}}}$$



UNNORMALIZED ANALOG TRANSFER FUNCTION



- 5. Determine the unnormalized analog transfer function H (s) is obtained by replacing s_n by s/Ω_c in the normalized transfer function of the low pass filter function
- When N is even,

$$\therefore H(s) = \prod_{k=1}^{\frac{N}{2}} \frac{1}{s_n^2 + b_k s_n + 1}$$

$$= \prod_{k=1}^{\frac{N}{2}} \frac{\Omega_c^2}{s^2 + b_k \Omega_c s + \Omega_c^2}$$



UNNORMALIZED ANALOG TRANSFER FUNCTION



- H(s) be the normalized Butterworth lowpass filter function
- When N is odd, H(s) is obtained by letting $s_n\to s/\Omega_c$ in the normalized Butterworth lowpass filter function

$$\therefore H(s) = \frac{1}{s_n + 1} \frac{\prod_{k=1}^{N-1}}{\prod_{k=1}^{2} \frac{1}{s_n^2 + b_k s_n + 1}}$$

$$= \frac{\Omega_c}{s + \Omega_c} \frac{\prod_{k=1}^{N-1}}{\prod_{k=1}^{2} \frac{\Omega_c^2}{s^2 + b_k \Omega_c + \Omega_c^2}}$$



DESIGN PROCEDURE FOR LOWPASS DIGITAL BUTTERWORTH IIR FILTER



- 6. Determine the transfer function of digital filter H(z). Using the suitable transformation to transform H(s) to H(z). When the Impulse invariant transformation is employed, if T<1, then multiply H(z) by T to normalize the magnitude.
- 7. Realize the digital filter transfer function H(z) by a suitable structure
- 8. Verify the design by sketching the frequency response H ($e^{j\omega}$)

$$H(e^{j\omega}) = H(z) / z = e^{j\omega}$$



ASSESSMENT



- 1. Compare Impulse Invariant and Bilinear transformation?
- 2. What is Butterworth approximation?
- 3. How will you choose the order N for a Butterworth Filter?
- 4. List the Properties of Butterworth Filter.
- 5. Analog filter transfer function is converted to a digital filter transfer function by using either ------ (or) ------
- 6. Define Sampling Time.





THANK YOU