

SNS COLLEGE OF TECHNOLOGY

Coimbatore – 35

An Autonomous Institution



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DEPARTMENT OF ELECTRONICS AND COMMUNICATION ENGINEERING

16EC401 / Wireless Communication

IV ECE/ VII SEMESTER

Unit IV - MULTIPATH MITIGATION TECHNIQUES

Topic : Error probability in fading channels With diversity reception



Introduction



Narrowband system:

Flat fading channel, single-tap channel model, performance enhancement through diversity





Introduction



Adjacent symbols (bits) do not affect the decision process (there is no intersymbol interference).







In a flat fading channel (or narrowband system), the CIR (channel impulse response) reduces to a single impulse scaled by a time-varying complex coefficient.

The received (equivalent lowpass) signal is of the form

 $r(t) = a(t)e^{j\phi(t)}s(t) + n(t)$

We assume that the phase changes "slowly" and can be perfectly tracked => important for coherent detection



Activity



≻Imagine folding a paper in half once

≻Then take the result and fold it in half again; and so on

≻How many times can you do that?



We assume:

The time-variant complex channel coefficient changes slowly (=> constant during a symbol interval)

The channel coefficient magnitude (= attenuation factor) a is a Rayleigh distributed random variable

Coherent detection of a binary PSK signal (assuming ideal phase synchronization)

Let us define instantaneous SNR and average SNR:

$$\gamma = a^2 E_b / N_0 \qquad \gamma_0 = E \left\{ a^2 \right\} \cdot E_b / N_0$$











Since





where the bit error probability for a certain value of *a* is

$$P_e(\gamma) = Q\left(\sqrt{2a^2E_b/N_0}\right) = Q\left(\sqrt{2\gamma}\right).$$

We thus get

$$P_e = \int_0^\infty Q\left(\sqrt{2\gamma}\right) \frac{1}{\gamma_0} e^{-\gamma/\gamma_0} d\gamma = \frac{1}{2} \left(1 - \sqrt{\frac{\gamma_0}{1 + \gamma_0}}\right).$$







Approximation for large values of average SNR is obtained in the following way. First, we write

$$P_{e} = \frac{1}{2} \left(1 - \sqrt{\frac{\gamma_{0}}{1 + \gamma_{0}}} \right) = \frac{1}{2} \left(1 - \sqrt{1 + \frac{-1}{1 + \gamma_{0}}} \right)$$

Then, we use

$$\sqrt{1+x} = 1 + x/2 + \dots$$

which leads to

$$P_e \approx 1/4\gamma_0$$
 for large γ_0 .



3/13/2024







Modulation
$$P_e(\gamma)$$
 P_e P_e (for large γ_0)2-PSK $Q(\sqrt{2\gamma})$ $\frac{1}{2}(1-\sqrt{\frac{\gamma_0}{1+\gamma_0}})$ $1/4\gamma_0$ DPSK $e^{-\gamma}/2$ $1/(2\gamma_0+2)$ $1/2\gamma_0$ 2-FSK
(coh.) $Q(\sqrt{\gamma})$ $\frac{1}{2}(1-\sqrt{\frac{\gamma_0}{2+\gamma_0}})$ $1/2\gamma_0$ 2-FSK
(non-c.) $e^{-\gamma/2}/2$ $1/(\gamma_0+2)$ $1/\gamma_0$



Assessment



> What are the modes of adaptive equalizer?

- a) Training mode
- b) Tracking mode
- c) Both of the mentioned
- d) None of the mentioned



> The ISI and adjacent channel interference is removed by

- a) Cancelling filter
- b) Port processing equalizer
- c) Both of the mentioned
- d) None of the mentioned





THANK YOU

Error in fading channels/16EC401 Wireless Communication /Dr.J.Geetha Ramani /ECE/SNSCT