



# SNS COLLEGE OF ENGINEERING



(Autonomous)

DEPARTMENT OF ELECTRONICS AND COMMUNICATION ENGINEERING

## UNIT- II

### IIR Filter Design

### Analog Filters-Butterworth filter



# ANALOGY FILTER DESIGN:

The most general form of analog filter transfer fn  $H(s)$ ,

$$H(s) = \frac{N(s)}{D(s)}$$

$$H(s) = \sum_{i=0}^M b_i s^i$$

$$H(s) = \sum_{i=1}^N a_i s^i$$

where,  $H(s)$  is the Laplace transform of the impulse response  $h(t)$ ,  
 i.e.,  $H(s) = \int_{-\infty}^{\infty} h(t) \cdot e^{-st} \cdot dt$

It should satisfy the condition,  
 $N \geq M$

For a stable analog filter the poles of  $H(s)$  should lie in left half of  $s$ -plane.

TYPES OF ANALOG FILTER:

- i) Butterworth filter
- ii) Chebyshev filter

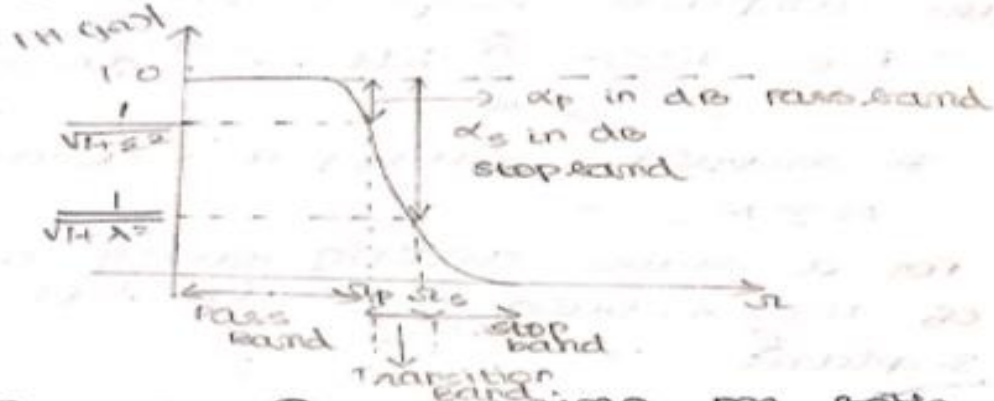
(ANALOG LOW PASS BUTTERWORTH FILTER)

Let the maximum pass band attenuation in possible decibels (dB) is  $\alpha_p (< 3 \text{ dB})$ . At pass band frequency  $\omega_p$  and  $\alpha_s$  in the minimum stop band attenuation in dB at stop band frequency  $\omega_s$ .

$$|H(j\omega)| = \frac{1}{\left[1 + \sum_{k=1}^N \left(\frac{\omega}{\omega_p}\right)^{2k}\right]^{1/2}} \quad (1)$$

↓  
epsilon

## Derivation Butterworth Approximation of Magnitude response :



From eqn ①, squaring on both sides,

$$|H(j\omega)|^2 = \frac{1}{\left[1 + \epsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N}\right]}$$

Taking log on both sides & multiplying by 10,

$$10 \log |H(j\omega)|^2 = 10 \log (1) - 10 \log \left[1 + \epsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N}\right]$$

$$20 \log |H(j\omega)| = (0) - 10 \log \left[1 + \epsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N}\right]$$

$$20 \log |H(j\omega)| = -10 \log \left[1 + \epsilon^2 \left(\frac{\omega}{\omega_p}\right)^{2N}\right] \quad \text{--- ②}$$

From figure at  $\omega = \omega_p$  the attenuation is equal to  $-\alpha_p$ .

$\therefore$  Eqn ②, becomes,

$$20 \log |H(j\omega_p)| = -10 \log \left[ 1 + \underbrace{\epsilon^2}_{\text{epsilon}} \left( \frac{\omega_p}{\omega_p} \right)^{2N} \right] = -\alpha_p$$

$$\alpha_p = 10 \log [1 + \epsilon^2]$$

$$0.1 \alpha_p = \log (1 + \epsilon^2)$$

$$10^{0.1 \alpha_p} = (1 + \epsilon^2)$$

$$\epsilon^2 = 10^{0.1 \alpha_p} - 1 \quad \text{--- (3)}$$

$$\epsilon = \sqrt{10^{0.1 \alpha_p} - 1} \quad \text{--- (3a)}$$

From figure at  $\omega = \omega_s$  the minimum stopband frequency is  $-\alpha_s$ .

Therefore eqn (3), becomes,

$$20 \log |H(j\omega_s)| = -\alpha_s = -10 \log \left[ 1 + \epsilon^2 \left( \frac{\omega_s}{\omega_p} \right)^{2N} \right]$$

$$\alpha_s = 10 \log \left[ 1 + \epsilon^2 \left( \frac{\omega_s}{\omega_p} \right)^{2N} \right]$$

$$0.1 \alpha_s = \log \left[ 1 + \epsilon^2 \left( \frac{\omega_s}{\omega_p} \right)^{2N} \right]$$

Taking antilog on both sides,

$$10^{0.1 \alpha_s} = \left[ 1 + \epsilon^2 \left( \frac{\omega_s}{\omega_p} \right)^{2N} \right]$$

From eqn (3a),

$$\epsilon^2 = 10^{0.1 \alpha_p} - 1$$

$$10^{0.1 \alpha_s} - 1 = (10^{0.1 \alpha_p} - 1) \left( \frac{\omega_s}{\omega_p} \right)^{2N}$$

$$\frac{10^{0.1 \alpha_s} - 1}{10^{0.1 \alpha_p} - 1} = \left( \frac{\omega_s}{\omega_p} \right)^{2N}$$



Taking  $\log$  on both sides,

$$\log \left[ \frac{10^{0.1ds} - 1}{10^{0.1dp} - 1} \right] = 2N \log \left( \frac{\omega_s}{\omega_p} \right)$$

$$2N = \log \left[ \frac{10^{0.1ds} - 1}{10^{0.1dp} - 1} \right]$$

$$\log \left( \frac{\omega_s}{\omega_p} \right)$$

$$N = \frac{\log \left[ \frac{10^{0.1ds} - 1}{10^{0.1dp} - 1} \right]}{2 \log \left( \frac{\omega_s}{\omega_p} \right)}$$

$$= \frac{1}{2} \log \left[ \frac{10^{0.1ds} - 1}{10^{0.1dp} - 1} \right]$$

$$\log \left( \frac{\omega_s}{\omega_p} \right)$$

$$= \frac{\log \left[ \frac{10^{0.1ds} - 1}{10^{0.1dp} - 1} \right]}{2 \log \left( \frac{\omega_s}{\omega_p} \right)}$$

$$N \approx \frac{\log \left[ \frac{10^{0.1ds} - 1}{10^{0.1dp} - 1} \right]}{2 \log \left( \frac{\omega_s}{\omega_p} \right)}$$

to round off the higher integer,  
i.e.,

$$N \geq \frac{\log \sqrt{\frac{10^{0.1d_s} - 1}{10^{0.1d_p} - 1}}}{\log \left[ \frac{\omega_s}{\omega_p} \right]}$$

$$N \geq \frac{\log \left[ \lambda / \epsilon \right]}{\log \left[ \frac{\omega_s}{\omega_p} \right]}$$

where,

$$\epsilon = \sqrt{10^{0.1d_p} - 1} \quad ; \quad \lambda = \sqrt{10^{0.1d_s} - 1}$$

$$A = \lambda / \epsilon \quad ; \quad k = \frac{\omega_p}{\omega_s} = \frac{1}{\frac{\omega_s}{\omega_p}}$$

where  $k \rightarrow$  transition ratio.

$N$  is order eqn for low pass Butterworth analog filter.

$$N \geq \frac{\log A}{\log \left( \frac{1}{k} \right)} \quad ; \quad N \geq \frac{\log A}{\log \left( \frac{\omega_s}{\omega_p} \right)}$$

Steps to design an analog Butterworth low pass filter :

STEP 1:

From the given specifications find the order of the filter.

$$- N \geq \frac{\log \sqrt{\frac{10^{0.1d_s} - 1}{10^{0.1d_p} - 1}}}{\log \left[ \frac{\omega_s}{\omega_p} \right]}$$

STEP 2:

Round off it to next higher integer.

STEP 3:

Find the transfer fn  $H(s)$

for  $\omega_c = 1 \text{ rad/sec}$  for the value of  $N$

calculate the value of cut-off frequency.

$$\omega_c = \frac{\omega_p}{(10^{0.1 \times N} - 1)^{1/2N}}$$

STEP 4:

Find the transfer fn  $H_d(s)$

for the above value of  $\omega_c$

by substituting

$$s \rightarrow \frac{s}{\omega_c} \text{ in } H(s)$$



ORDER N	Denominator of H(s) for $\omega_c = 1 \text{ rad/sec}$
1	$s+1$
2	$s^2 + \sqrt{2}s + 1$
3	$(s+1) \cdot (s^2 + s + 1)$
4	$(s^2 + 0.76537s + 1) (s^2 + 1.84775s + 1)$
5	$(s+1) (s^2 + 0.61803s + 1) (s^2 + 1.61803s + 1)$
6	$(s^2 + 1.931855s + 1) (s^2 + \sqrt{2}s + 1) (s^2 + 0.51764s + 1)$



For the given specification design an Analog Butterworth filter  $\alpha_p = 1 \text{ dB}$ ,  $\alpha_s = 30 \text{ dB}$ .  $\omega_p = 200 \text{ rad/sec}$ ,  $\omega_s = 600 \text{ rad/sec}$

Sol:

STEP 1:

$$N \geq \frac{\log \sqrt{\frac{10^{0.1\alpha_s} - 1}{10^{0.1\alpha_p} - 1}}}{\log \left( \frac{\omega_s}{\omega_p} \right)}$$

$$N \geq \frac{\log \sqrt{\frac{10^{0.1 \times 30} - 1}{10^{0.1 \times 1} - 1}}}{\log \left( \frac{600}{200} \right)}$$

$$N \geq \frac{\log (62.1148)}{\log 3}$$

$$N \geq \frac{1.793}{0.477}$$

$$N \geq 3.75$$

STEP 2:

$$N \geq 3.75$$

$$N = 4$$

STEP-3:

$$H(s) = \frac{1}{(s^2 + 0.7653s + 1)(s^2 + 1.8477s + 1)}$$

calculated the value for  $\omega_c$  (1)

$$\begin{aligned} \omega_c &= \frac{\omega_p}{(10^{0.1K_p} - 1)^{1/2N}} = \frac{200}{(10^{(0.1 \times 1)} - 1)^{1/2 \times 4}} \\ &= \frac{200}{(0.258)^{1/8}} = \frac{200}{(0.258)^{0.125}} \end{aligned}$$

$$\omega_c = 236.90 \text{ rad/sec}$$

STEP-4:

$$s \rightarrow \frac{s}{\omega_c} = \frac{s}{236.90} \quad 0.8442$$

sub  $s \rightarrow \frac{s}{236.90}$  in Eqn (1)

$$H_a(s) = \frac{1}{\left[ \left( \frac{s}{236.90} \right)^2 + 0.7653 \left( \frac{s}{236.90} \right) + 1 \right] \left[ \left( \frac{s}{236.90} \right)^2 + 1.8477 \left( \frac{s}{236.90} \right) + 1 \right]}$$

$$= \frac{1}{\left[ (1.78 \times 10^{-5} s)^2 + (3.23 \times 10^{-3} s) + 1 \right] \left[ (1.78 \times 10^{-5} s)^2 + (7.79 \times 10^{-3} s) + 1 \right]}$$

$$= \frac{1}{(1.78 \times 10^{-5})^2 \left[ s^2 + \frac{3.23 \times 10^{-3} s}{1.78 \times 10^{-5}} + \frac{1}{1.78 \times 10^{-5}} \right] \left[ s^2 + \frac{7.79 \times 10^{-3} s}{1.78 \times 10^{-5}} + \frac{1}{1.78 \times 10^{-5}} \right]}$$

$$= \frac{1}{\left[ s^2 + 181.46s + 56 \times 10^3 \right] \left[ s^2 + 437.6s + 56 \times 10^3 \right]}$$



Thank You!